SURFACE WAVE PROPAGATION THROUGH ANISOTROPIC CIRCULAR, CYLINDRICAL DIELECTRIC ROD WAVEGUIDE*

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ABSTRACT

The expressions for the propagation characteristics, such as, the radial propagation constants (K), the axial propagation constant (γ), the guided wavelength (λ_g), and the cut-off conditions for a circular, cylindrical, anisotropic dielectric rod waveguide excited by the symmetric electric (TM_{01}) and symmetric magnetic (TE_{01}) modes have been derived. Numerical calculation of these characteristics (K, γ, λ_g) as a function of the diameter (d) and the dielectri constant (ϵ_1) of the guide have been completed and the results are presented graphically. The electric and the magnetic field distributions inside and outside the waveguide as a function of a diameter (d) are also given.

1. INTROUCTION

The propagation characteristics of surface waves on different structures have been discussed in an illuminating manner by Barlow and Brown (1962)

and others. Several authors have studied the surface wave structures in the form of corrugated, circular, cylindrical guide. Barlow and Karbowiak (1954) did the experimental investigations of surface waves on corrugated, circular, cylindrical metal rods. Simon and Weill (1952) have used a limiting form of the corrugated conducting cylinder excited in the dipole mode, consisting of an array of conducting disks normal to the axis and spaced at regular, periodic intervals. Piefke (1959) has studied the corrugated structure excited in the HE_{11} mode. The object of this paper is to present the theoretical investigations on the propagation of microwaves through an anisotropic dielectric rod waveguide excited in TE_{01} and TM_{01} modes

2. FIELD COMPONENTS

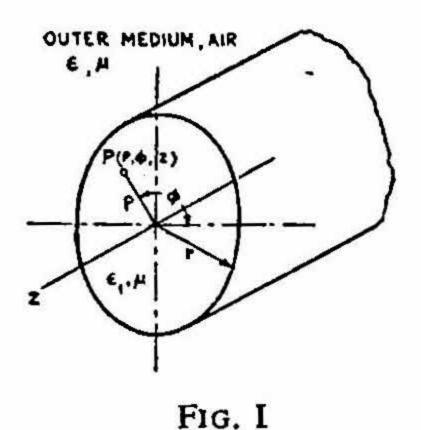
The field components inside and outside the waveguide (Figure I) when excited in the TE_{01} and TM_{01} modes have been derived from the Maxwell's

^{*} My attention has been drawn to a previous paper on the same subject by P. R. Longaker appearing in IEEE Transactions on MTT, Vol. 11, November 1963. I wish to state that I was aware of this particular paper in question and I have corresponded with Dr. Longaker wherein I had pointed out certain discrepancies in his paper, which have been acknowledged by him. In view of this, this previous publication does not in any way detract the value of this communication.

Equations by solving the wave equation and are given below. A harmonic time dependence of the field components has been assumed and is omitted in the expressions for the sake of convenience. The anisotropic dielectric rod waveguide and the surrounding medium (Figure 1), are assumed to be lossless and have the same permeability, μ . Therefore, the propagation constant, $\gamma - i\beta = i2\pi/\lambda_g$, λ_g being the wavelength of the guided wave. The dielectric constant, ϵ_1 , of the waveguide is a tensor and is represented as follows,

$$\epsilon_{1} = \begin{cases} \epsilon_{p} & 0 & 0 \\ 0 & \epsilon_{p} & 0 \\ 0 & 0 & \epsilon_{z} \end{cases}$$

$$[1]$$



The Co-ordinate system (P, ϕ , z) in the anisotropic dielectric rod wave-guide

 $TE_{01} - Mode$ 2.1

Inside the waveguide, $\rho < r$, radius of the guide,

$$H_{z1} = A J_{o} \left(K_{1}^{E} \rho \right) \exp \left(-\gamma_{1}^{E} z \right)$$

$$H_{\rho 1} = iA \left[\overline{\epsilon}_{\rho} \left(\lambda_{g} / \lambda_{0} \right)^{2} - 1 \right]^{-1} J_{1} \left(K_{1}^{E} \rho \right), \exp \left(-\gamma_{1}^{E} z \right)$$

$$E_{\phi 1} = -iA \left[\mu / \epsilon \right]^{1} \left[\overline{\epsilon}_{\rho} - \left(\lambda_{0} / \lambda_{*} \right)^{2} \right]^{-1} J_{1} \left(K_{1}^{E} \rho \right), \exp \left(-\gamma_{1}^{E} z \right)$$

$$\left[2 \right]$$

Outside the waveguide, $\rho > r$,

$$H_{z2} = BH_0^{(1)} \left(K_2^E \rho\right) \cdot \exp\left(-\gamma_2^E z\right)$$

$$H_{\rho 2} = iB \left[\lambda_g / \lambda_0\right]^2 - 1 \right]^{-\frac{1}{2}} H_1^{(1)} \left(K_2^E \rho\right) \cdot \exp\left(-\gamma_2^E z\right) \qquad [3]$$

$$E_{\phi 2} = -iB \left[\mu / \epsilon\right]^{\frac{1}{2}} \left[1 - (\lambda_0 / \lambda_g)^2\right]^{-\frac{1}{2}} H_1^{(1)} \left(K_2^E \rho\right) \cdot \exp\left(-\gamma_2^E z\right)$$
The variation of γ_1^E and γ_2^E in the axial direction is considered as

$$\exp. (-\gamma^E z).$$

 $TM_{01} - Mode$ 2.2

Inside the waveguide, $\rho < r$,

$$E_{z1} = C J_0 (K_1^H \rho) \cdot \exp(-\gamma_1^H z)$$

$$E\rho_1 = iC \left[\overline{\epsilon_z}/\overline{\epsilon_\rho}\right] \left[\overline{\epsilon_z} (\lambda_g/\lambda_0)^2 - \overline{\epsilon_z}/\overline{\epsilon_\rho}\right]^{-\frac{1}{2}} J_1 (K_1^H \rho) \cdot \exp(-\gamma_1^H z) \quad [4]$$

$$H_{\phi I} = iC \overline{\epsilon_z} \left[\epsilon/\mu\right]^{\frac{1}{2}} \left[\overline{\epsilon_z} - (\lambda_0/\lambda_g)^2 \cdot (\overline{\epsilon_z}/\overline{\epsilon_\rho})\right]^{-\frac{1}{2}} J_1 (K_1^H \rho) \cdot \exp(-\gamma_1^H z)$$

Outside the waveguide, $\rho > r$,

$$E_{z2} = DH_0^{(1)} (K_2^H \rho) \cdot \exp \left(-\gamma_2^H z \right)$$

$$E_{\rho z} = i D \left[(\lambda_g / \lambda_0)^2 - 1 \right]^{-1} H_1^{(1)} (K_2^H \rho) \cdot \exp \left(-\gamma_2^H z \right)$$
[5]

$$H_{\phi 2} = i D \left[\epsilon / \mu \right]^{4} \left[1 - (\lambda_{0} / \lambda_{g})^{2} \right]^{-4} H_{1}^{(1)} \left(K_{2}^{H} \rho \right) \cdot \exp \left(- \gamma_{2}^{H} z \right)$$

The variation of γ_1^H and γ_2^H in the axial direction is considered as exp. $(-\gamma^{H}z)$.

In the equations (2) to (5)

 K_1^E = Radial propagation constant inside the guide for TE mode $= \left[\left(\sqrt{E} \right)^2 + \omega^2 \mu \epsilon_0 \right]^{\frac{1}{2}} = \left(2\pi / \lambda_0 \right) \left[\epsilon_P - \left(\lambda_0 / \lambda_0 \right)^2 \right]^{\frac{1}{2}}$

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$$= \left[\left(\gamma_1^E \right)^2 + \omega^2 \mu \epsilon_\rho \right]^{\dagger} = \left(2\pi / \lambda_0 \right) \left[\overline{\epsilon_\rho} - \left(\lambda_0 / \lambda_g \right)^2 \right]^{\dagger}$$
 [6]

 K_{i}^{H} = Radial propagation constant inside the guide for TM mode $= \left[\left\{ \gamma_1^H \right\}^2 + \omega^2 \mu \epsilon_p \right] \left(\overline{\epsilon_z} / \overline{\epsilon_p} \right) \right]^{\frac{1}{2}} = \left(2\pi / \lambda_0 \right) \left[\overline{\epsilon_z} - \left(\lambda_0 / \lambda_g \right)^2 \left(\overline{\epsilon_z} / \overline{\epsilon_p} \right) \right]^{\frac{1}{2}}$ [7]

$K_2 =$ Radial propagation constant outside the waveguide

$$= [\gamma^2 + \omega^2 \mu \epsilon]^{i} - (2\pi/\lambda_0) [1 - (\lambda_0/\lambda_g)^2]^{i}$$
[8]

$$\overline{\epsilon}_{z} = [\epsilon_{z}/\epsilon], \ \overline{\epsilon}_{p} = [\epsilon_{p}/\epsilon]$$
[9]

CHARACTERISTIC EQUATIONS 3.

The characteristic equations for both the modes as obtained by matching the field components at the boundary surface, $\rho = r$, are given as follows:

TE01 Mode 3.1

$$x_1 \frac{J_0(x_1)}{J_1(x_1)} = x_2 \frac{H_0^{(1)}(x_2)}{H_1^{(1)}(x_2)}$$
[10]

[11] Where, $x_1 = K_1^E r$ and $x_2 = K_2^E r$

From equations (6) and (8), we have

$$[x_1^2 + (x_2/i)^2] = (\pi d/\lambda_0)^2 [\bar{e}_p - 1]$$
 [12]

3.2 TM of Mode

$$y_1 \frac{J_0(y_1)}{J_1(y_1)} = \overline{\epsilon_z} y_2 \frac{H_0^{(1)}(y_2)}{H_1^{(1)}(y_2)}$$
[13]

where,
$$y_1 = K_1^H r$$
 and $y_2 = K_2^H r$ [14]

From equations [7] and [8], we have,

$$\left[y^2 + y_2/i\right)^2 \left(\overline{\epsilon_z}/\overline{\epsilon_p}\right) = \left(\pi d/\lambda_0\right)^2 \left[\overline{\epsilon_z} - \left(\overline{\epsilon_z}/\overline{\epsilon_p}\right)\right]$$
[15]

4. SOLUTION OF THE CHARACTERISTIC EQUATIONS

In solving these equations, we assume that there is no dielectric loss. Thus, γ^2 and hence K_1^2 will be real. To have a non-radiating mode, the arguments of the Hankel functions must be imaginary. Therefore, K_2 must be imaginary and so, K_1 is real. Now, given that K_1 is real and K_2 is imaginary, the characteristic equations are solved graphically.

4.1 Guided Wavelength

The expressions for the guided wavelength can be determined from the equations [6] and [8], and [7] and [8] for the TE and TM waves, respectively, and are given as follows:

(a) TE_{01} Mode:

From equations [6] and [8], we have on putting $x_1 = K_1^E r$ and $x_2 = K_2^E r$,

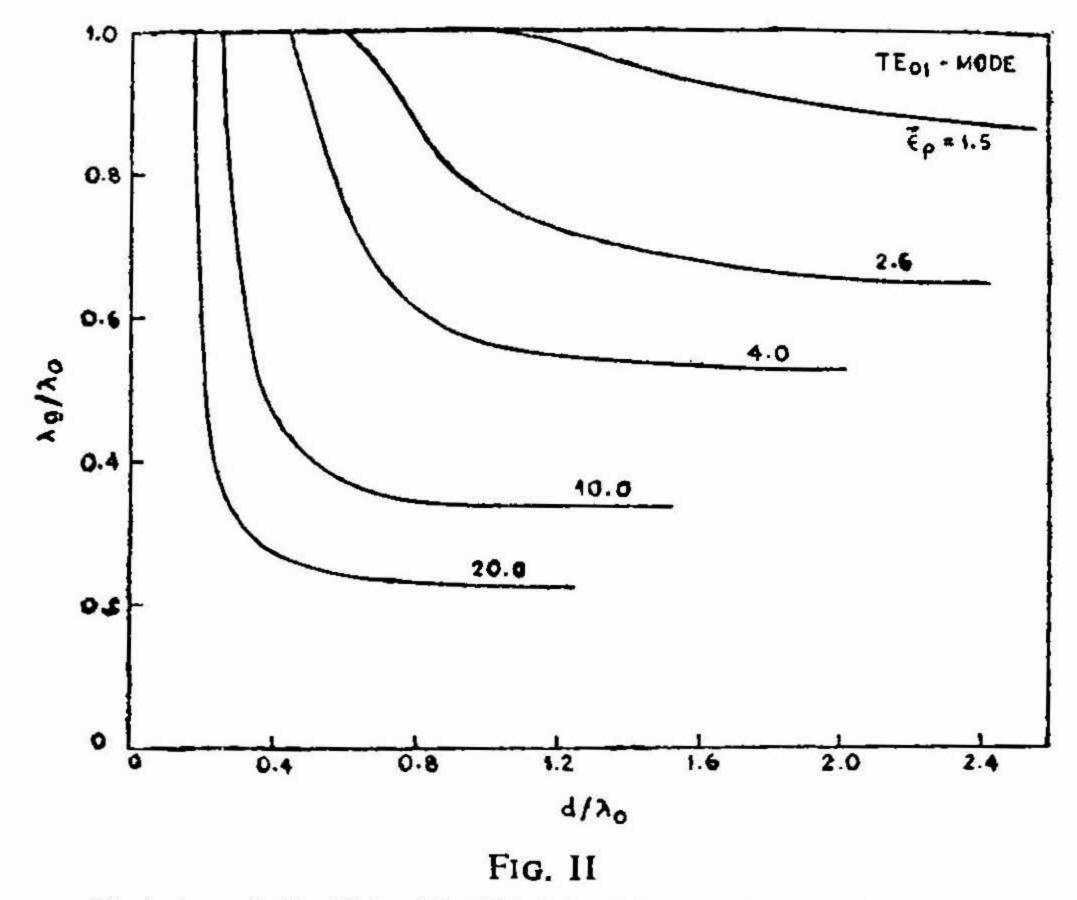
$$[\lambda_g/\lambda_0]^2 = [1 - (x_2 \lambda_0/\pi d)^2]^{-1} = [\overline{\epsilon_\rho} - (x_1 \lambda_0/\pi d)^2]^{-1}$$
[16]

The results calculated as a function of $(1/\lambda_0)$ and $\overline{\epsilon_{\rho}}$ are shown graphically in Figure II and Figure III, respectively.

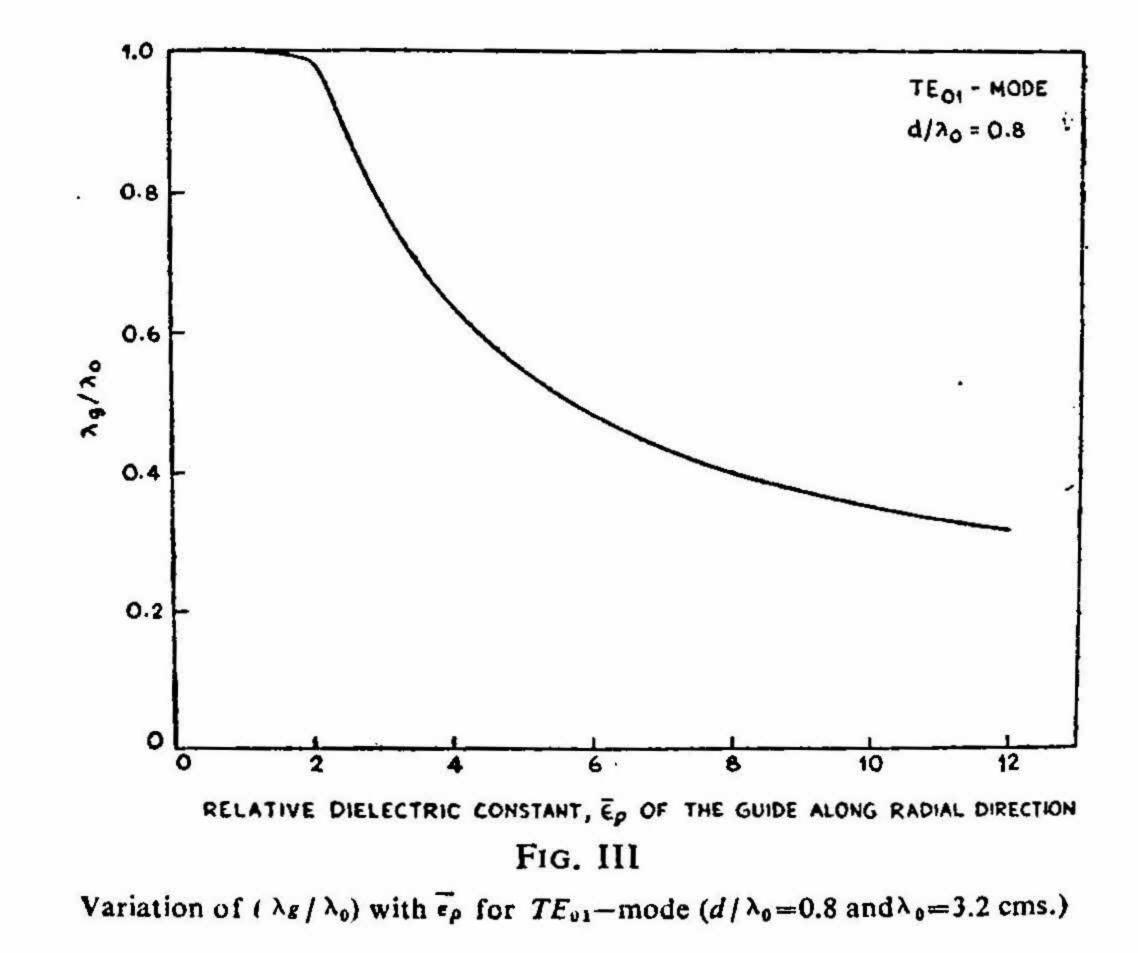
(b) TM_{01} Mode:

From equations [7] and [8], we have on putting $y_1 = K_1^H r$ and $y_2 = K_2^H r$, $[\lambda_g/\lambda_0]^2 = [1 - (y_2 \lambda_0/\pi d)^2]^{-1}$ $= [\overline{\epsilon_x}/\overline{\epsilon_y}] [\overline{\epsilon_x} - (y_1 \lambda_0/\pi d)^2]^{-1}$ [17]

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Variation of (λ_g / λ_0) with (d / λ_0) for TE_{01} -mode $(\lambda_0 = 3.2 \text{ cms.})$



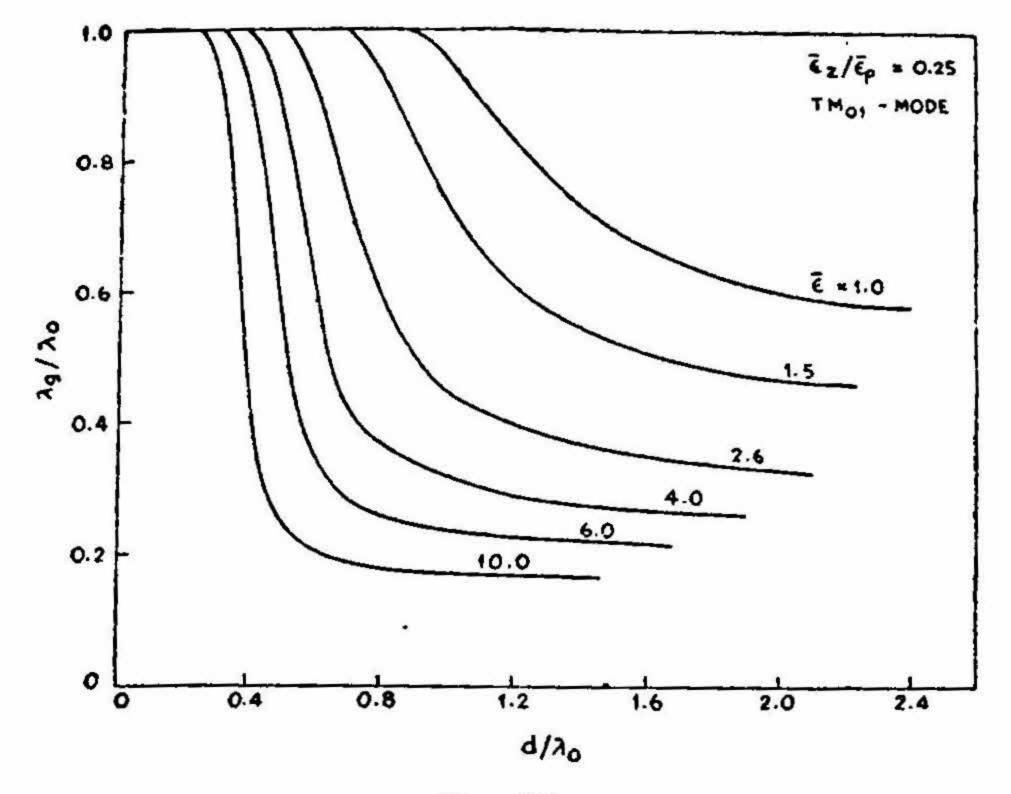
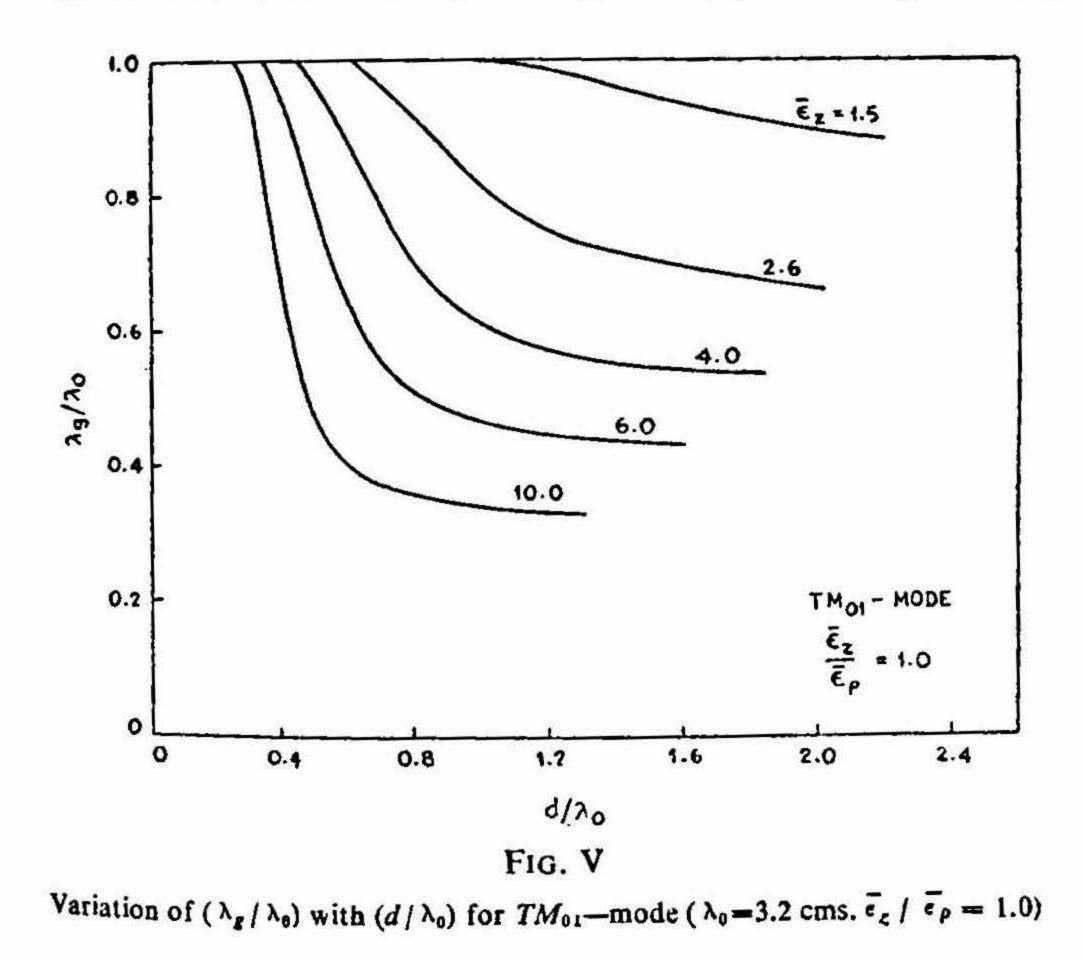


FIG. IV

Variation of $(\lambda_{\epsilon}/\lambda_{0})$ with (d/λ_{0}) for TM_{01} -mode $(\lambda_{0}=3.2 \text{ cms. } \overline{e_{\epsilon}}/\overline{e_{\rho}}=0.25)$



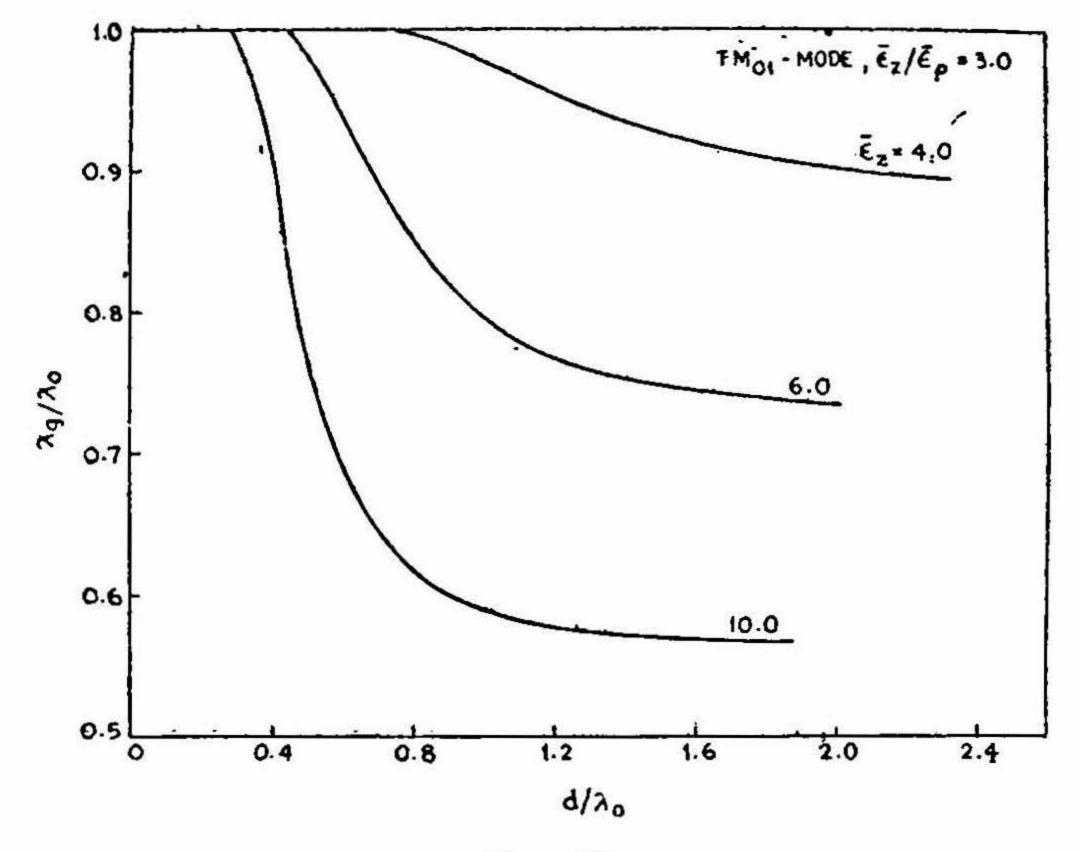
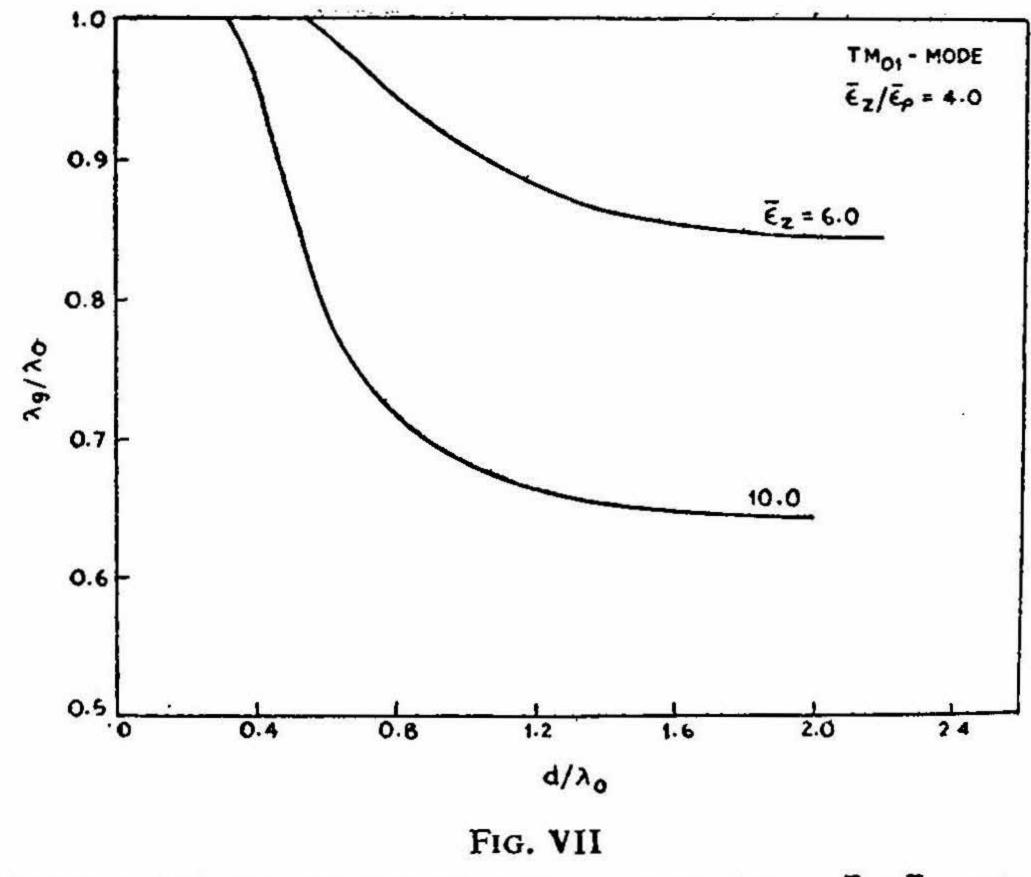


FIG. VI

Variation of (λ_g / λ_0) with (d / λ_0) for $\dot{T}M_{01}$ -mode $(\lambda_0 = 3.2 \text{ cms. } \overline{\epsilon_z} / \overline{\epsilon_\rho} = 3.0)$



Variation of (λ_g/λ_0) with (d/λ_0) for TM_{01} -mode $(\lambda=3.2 \text{ cms.} \overline{e_z}/\overline{e_p}=4.0)$

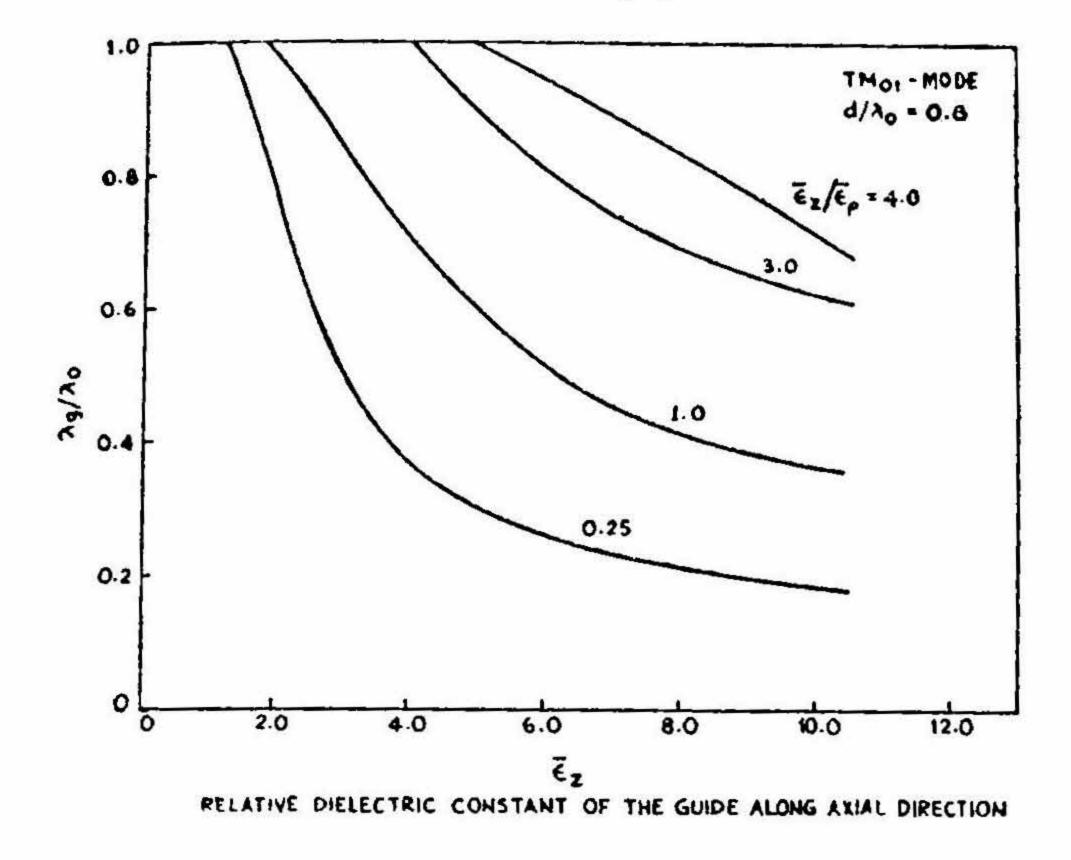


FIG. VIII

Variation of (λ_g / λ_0) with ϵ_z for TM_{01} -mode $(d / \lambda_0 = 0.8, \lambda_0 = 3.2$ cms.)

These are calculated as a function of (d/λ_0) and ϵ_z for different values of

 $[\epsilon_z/\epsilon_\rho]$ ranging from 0.25 to 4.0 and are shown graphically in Figures IV, V, VI, VII and VIII.

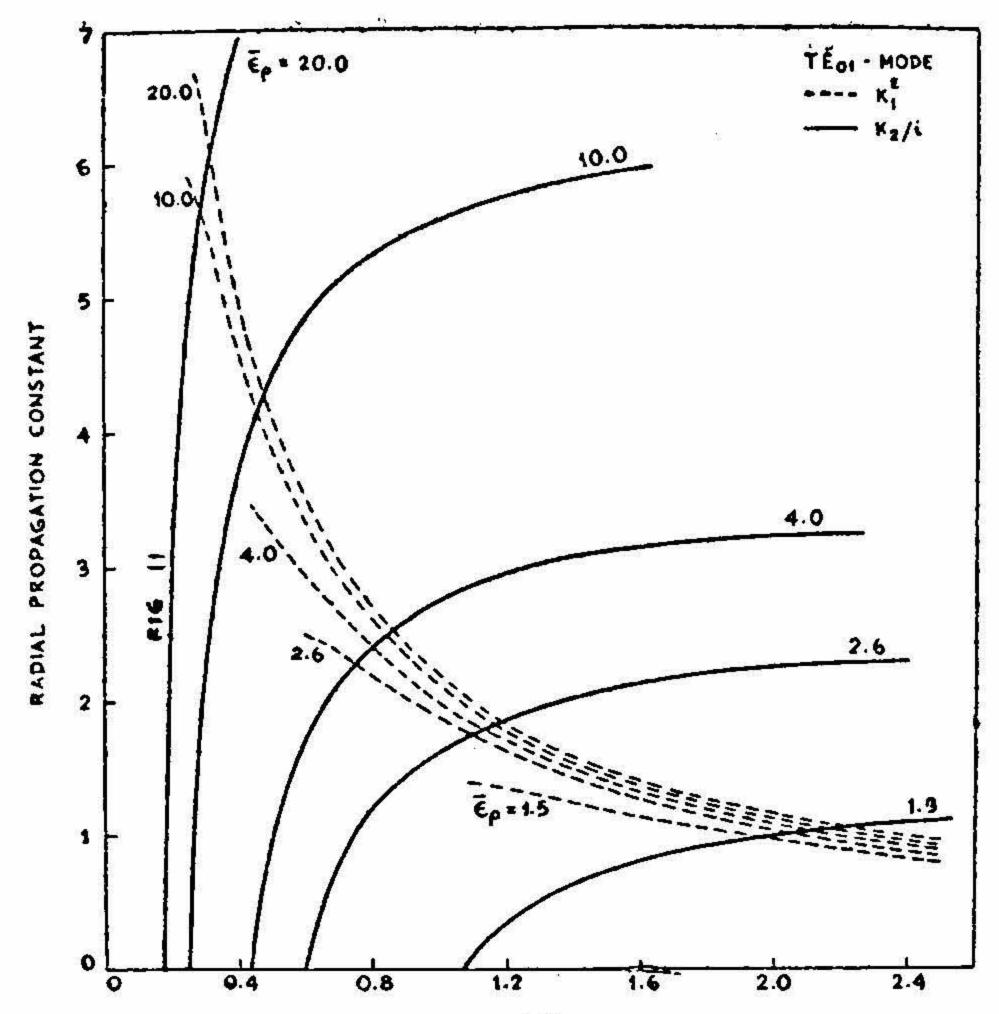
It will be seen, however, that for a dielectric medium of infinite extent, that is, (d/λ_0) tending to infinity, we have in the limiting case, $(\lambda_g/\lambda_0) = [\overline{\epsilon_p}]^{-\frac{1}{2}}$.

4.2 Radial Propagation Constants

The radial propagation constants $(K_1 \text{ and } K_2)$ are calculated from equations [11] and [14] for TE_{c_1} and TM_{01} modes, respectively. The variation of K_1 and K_2 as a function of (d/λ_0) for $\epsilon_p = 1.5$, 2.6, 4.0, 10.0, 20.0 are plotted in Fig. IX and as a function of ϵ_p with $(d/\lambda_0) = 0.8$ in Figure X for TE_{01} mode. For TM_{01} mode, these are plotted as a function of (d/λ_0) for $\epsilon_z = 1.0$, 1.5, 2.6, 4.0, 6.0, 10.0 with $[\epsilon_z/\epsilon_p]$ ranging from 0.25 to 4.0 in Figure XI, XII XIII and XIV and as a function of ϵ_z for $(d/\lambda_0) = 0.8$ with $[\epsilon_z/\epsilon_p]$ ranging from 0.25 to 4.0 in Figure XV.

4.3 Axial Propagation Constants

The axial propagation constants (γ) are calculated from equations [6] and [7] for TE_{01} and TM_{01} modes, respectively. The variation of (γ/γ_0) with



d/20

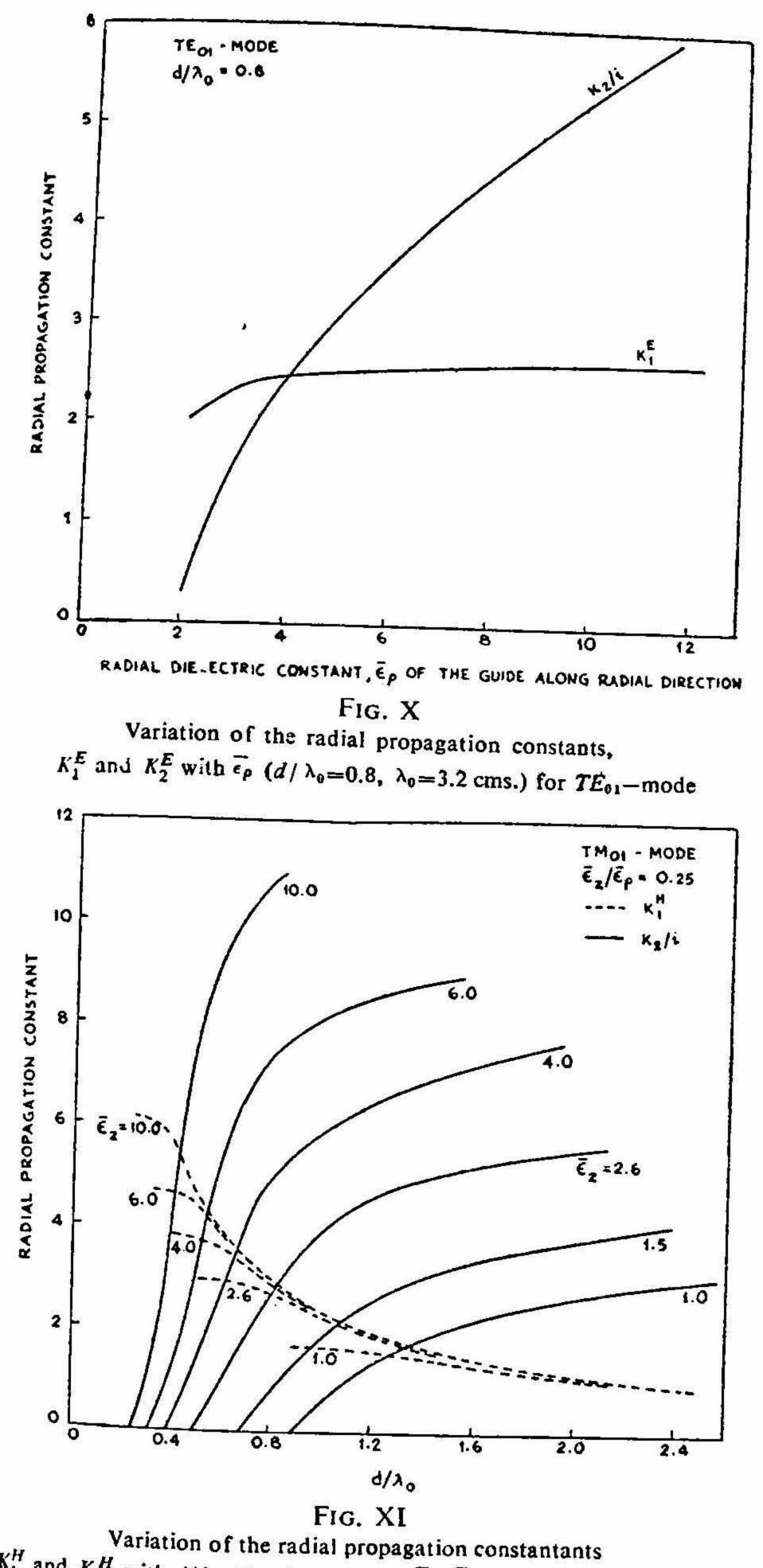
FIG. IX

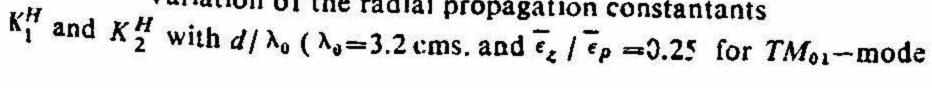
Variation of the radial propagation constants, K_1^E and K_2^E with d/λ_0 . ($\lambda_0 = 3.2$ cms.) for TE_{01} mode (As K, is imaginary K_2/i has been plotted)

respect to (d/λ_0) and ϵ_{ρ} for TE_{01} mode are shown in Figures XVI and XVII, respectively. For TM_{01} mode, these are plotted as a function of (d/λ_0) for $\epsilon_z = 1.0, 1.5, 2.6, 4.0, 6.0, 10.0$ with $[\epsilon_z/\epsilon_{\rho}]$ ranging from 0.25 to 4.0 in Figures XVIII, XIX, XX and XXI and as a function of ϵ_z for $(d/\lambda_0) = 0.8$ with $[\epsilon_z/\epsilon_{\rho}]$ ranging from 0.25 to 4.0 in Figure XXII.

4.4 Field Configurations

The electric and the magnetic field distributions inside and outside the anisotropic dielectric rod waveguide have been calculated from equations [2], [3], [4] and [5] for $(d/\lambda_0) = 0.8$. These are shown in in Figures XXIII, XXIV, XXV and in Figures XXVI, XXVII, XXVIII for TE_{01} and TM_{01} modes, respectively. The values of the field components have been normalised with respect to their values at the surface, $\rho = r$, of the guide.





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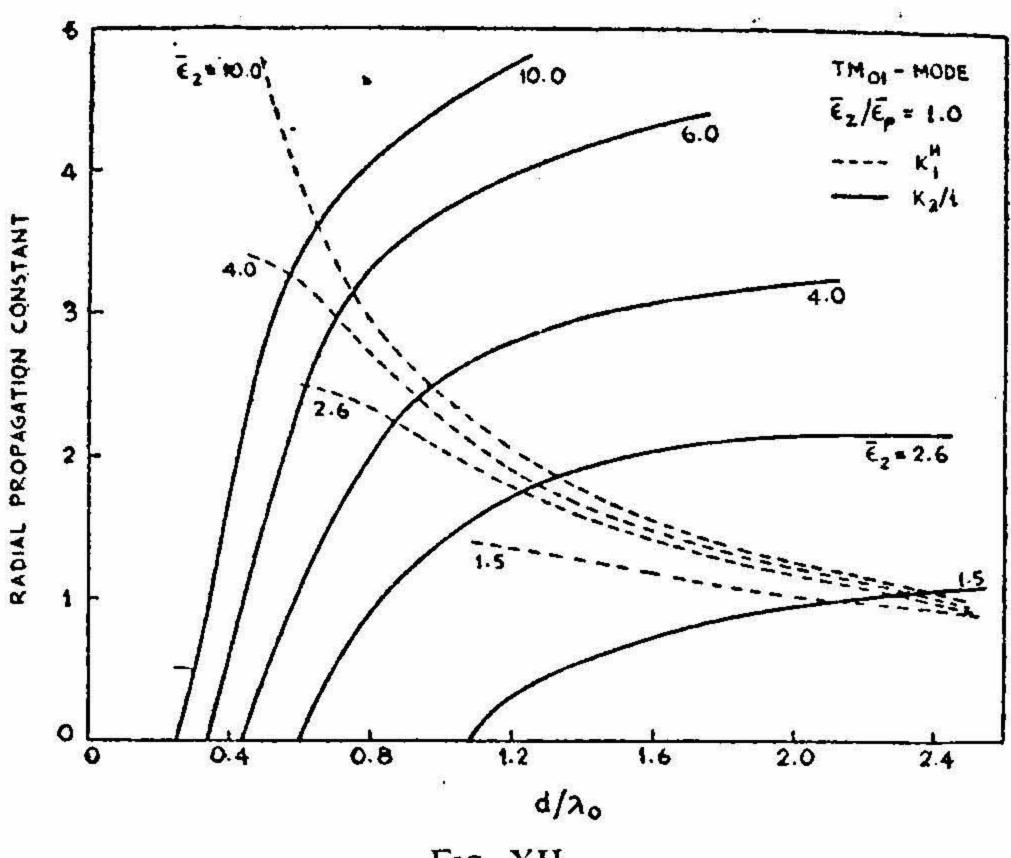


FIG. XII Variation of the radial propagation constants, K_1^H and K_2^H with d/λ_0 ($\lambda_0 = 3.2$ cms. $\overline{e_z}/\overline{e_p} = 1.0$) for TM_{01} -mode

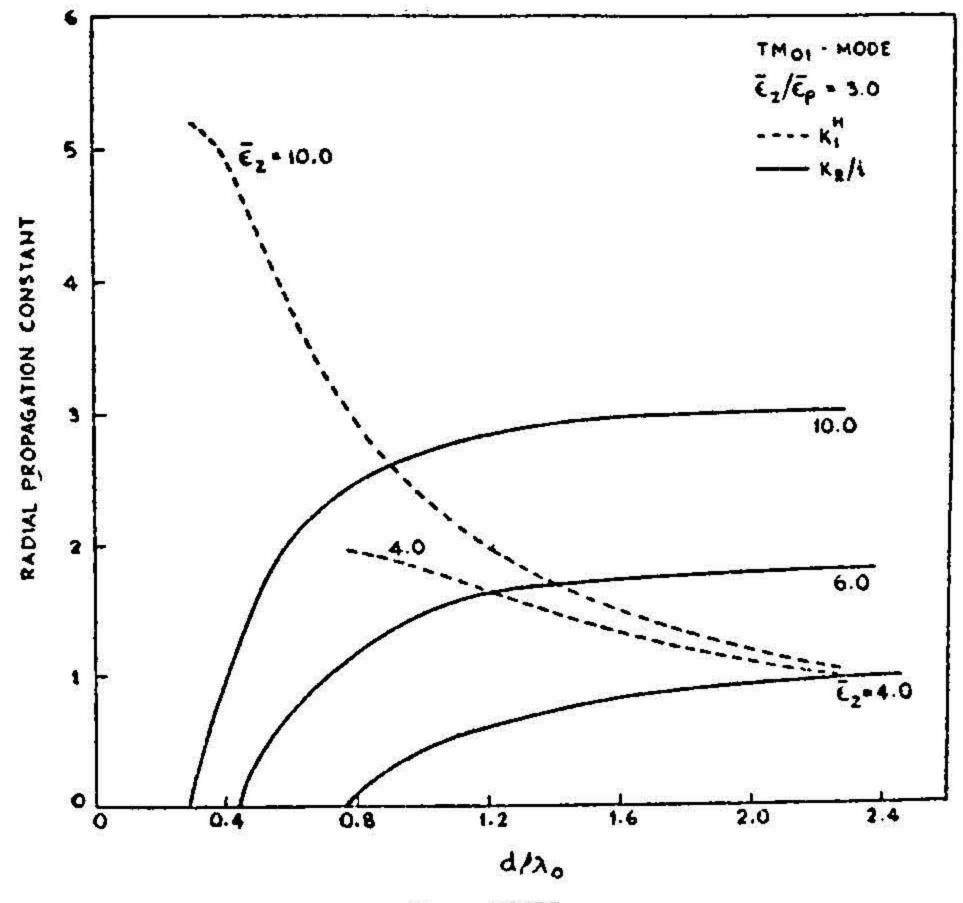
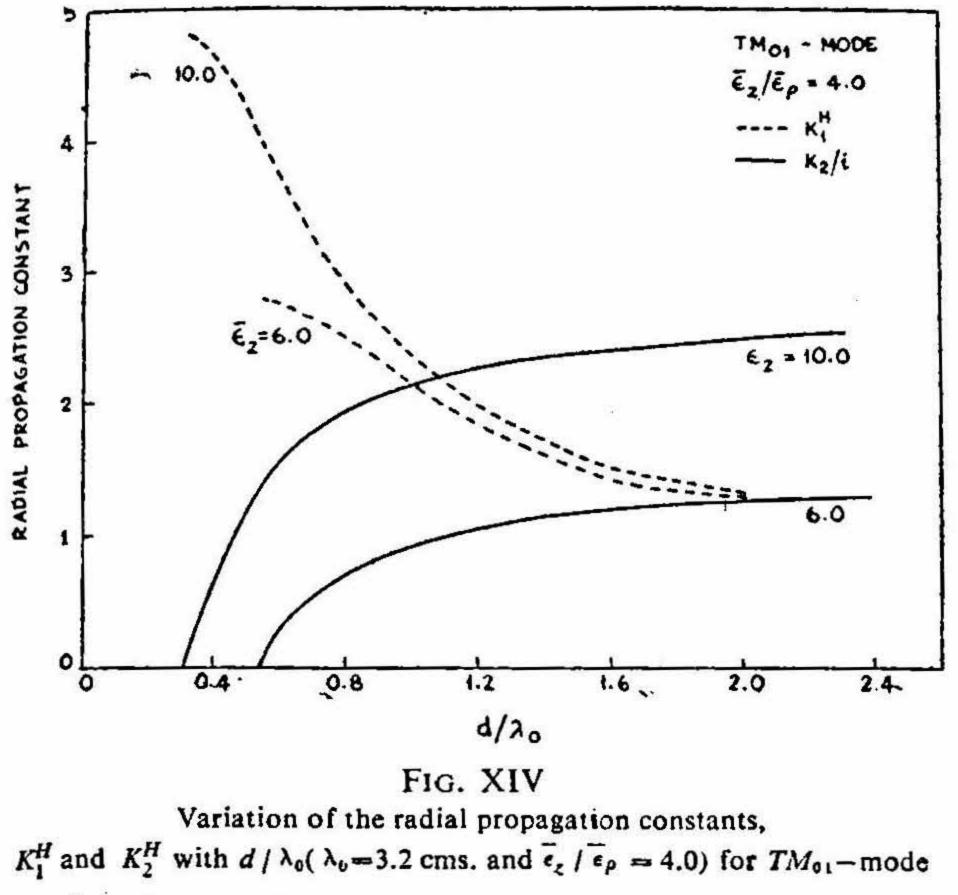
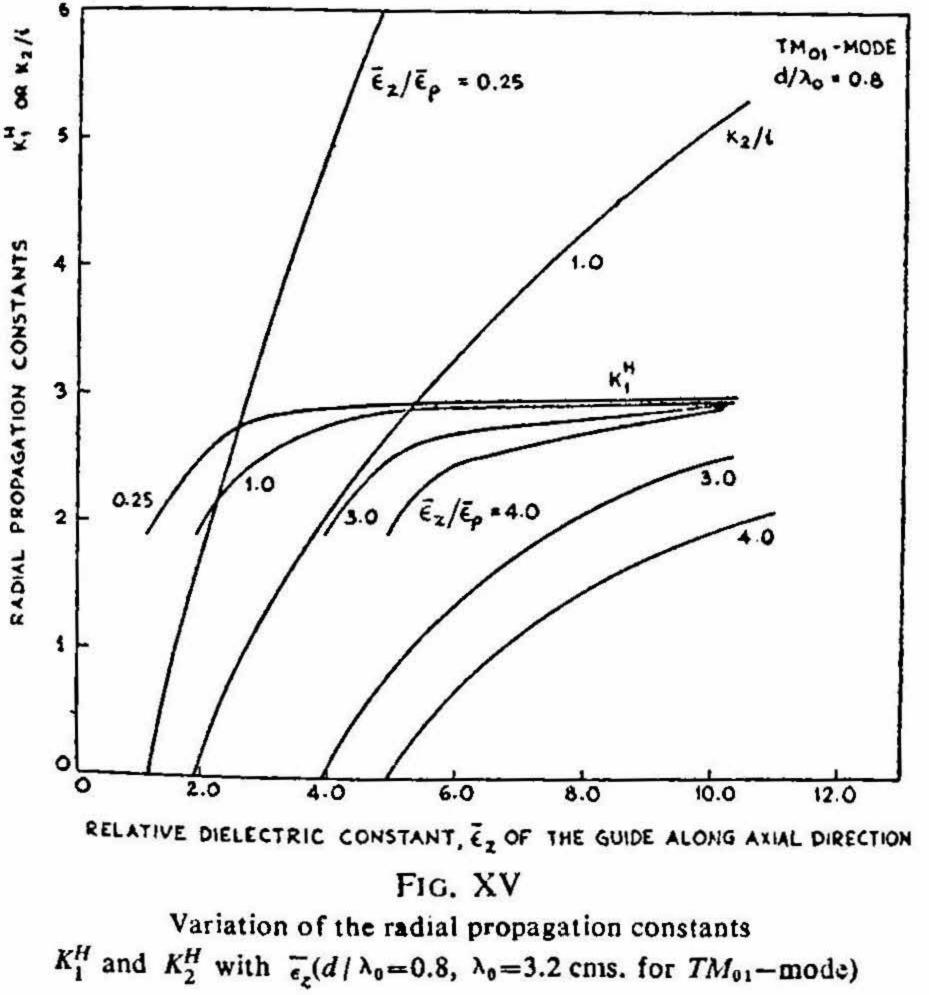
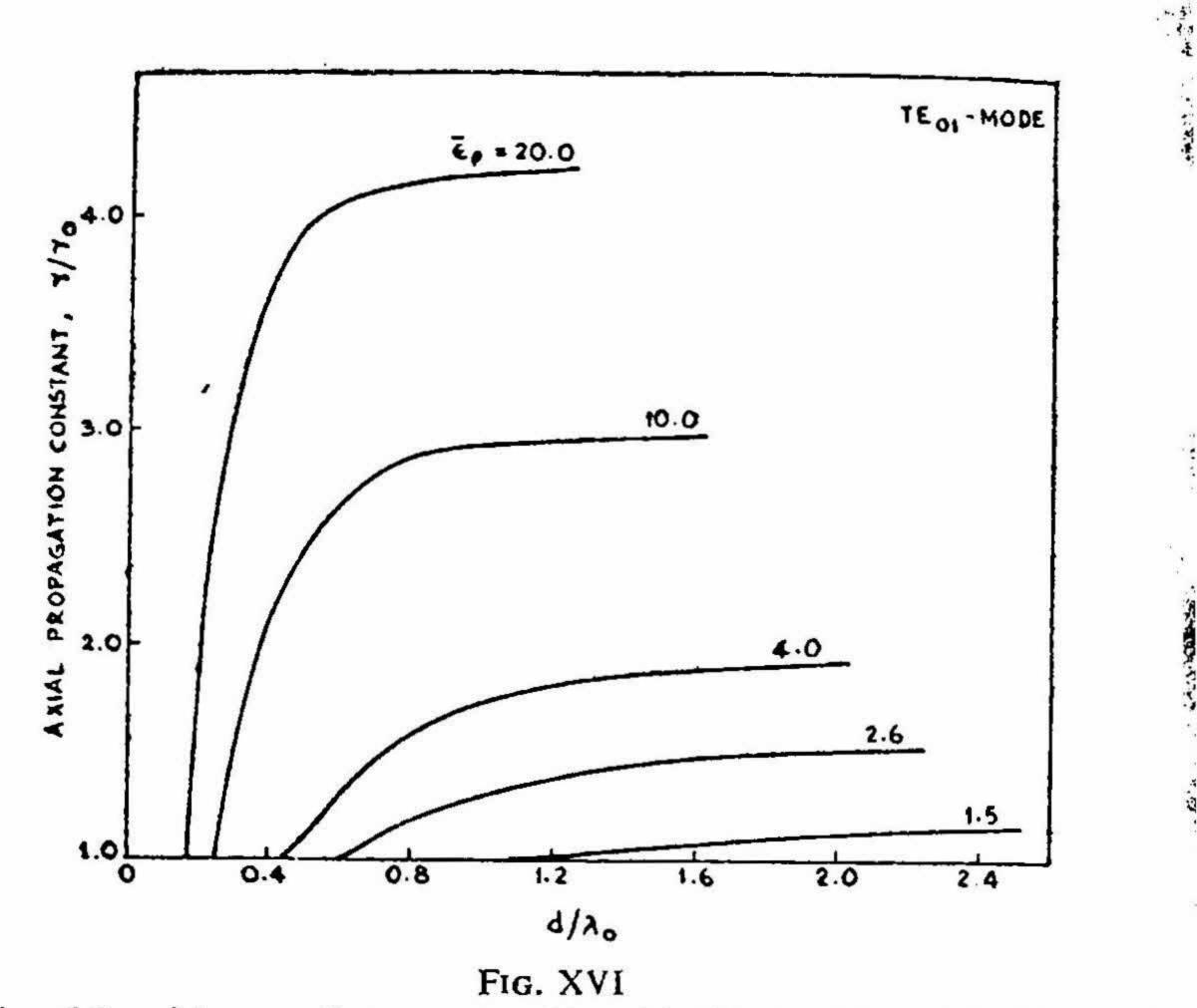


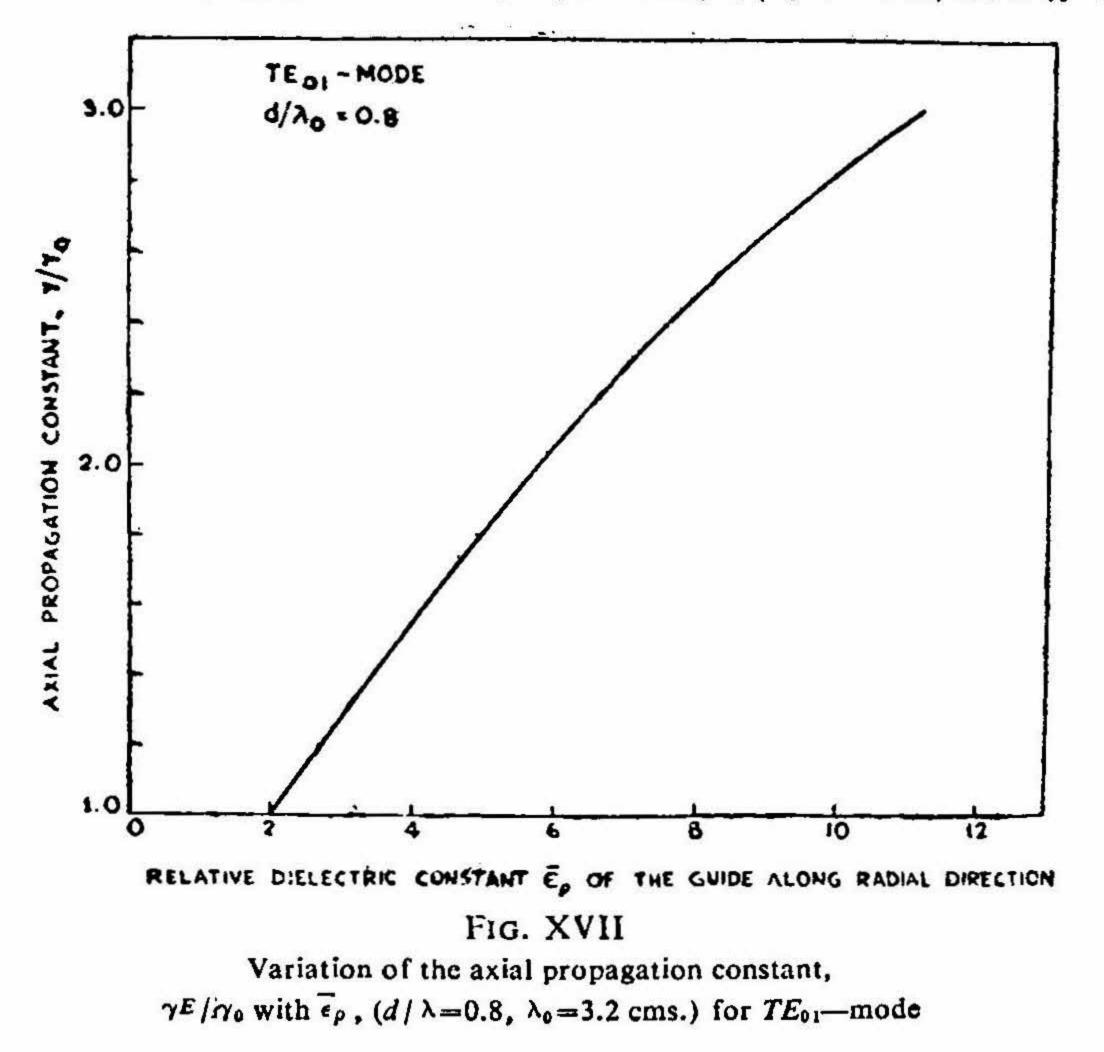
FIG. XIII Variation of the radial propagation constants, K_1^H and K_2^H with d/λ_0 ($\lambda_0 = 3.2 \text{ cms.} \overline{\epsilon_z} / \overline{\epsilon_\rho} = 3.0$) for TM_{01} -mode







Variation of the axial propagation constants, $\gamma E / \gamma_0$ with d / λ_0 ($\lambda_0 = 3.2$ cms.) for TE_{01} -mode

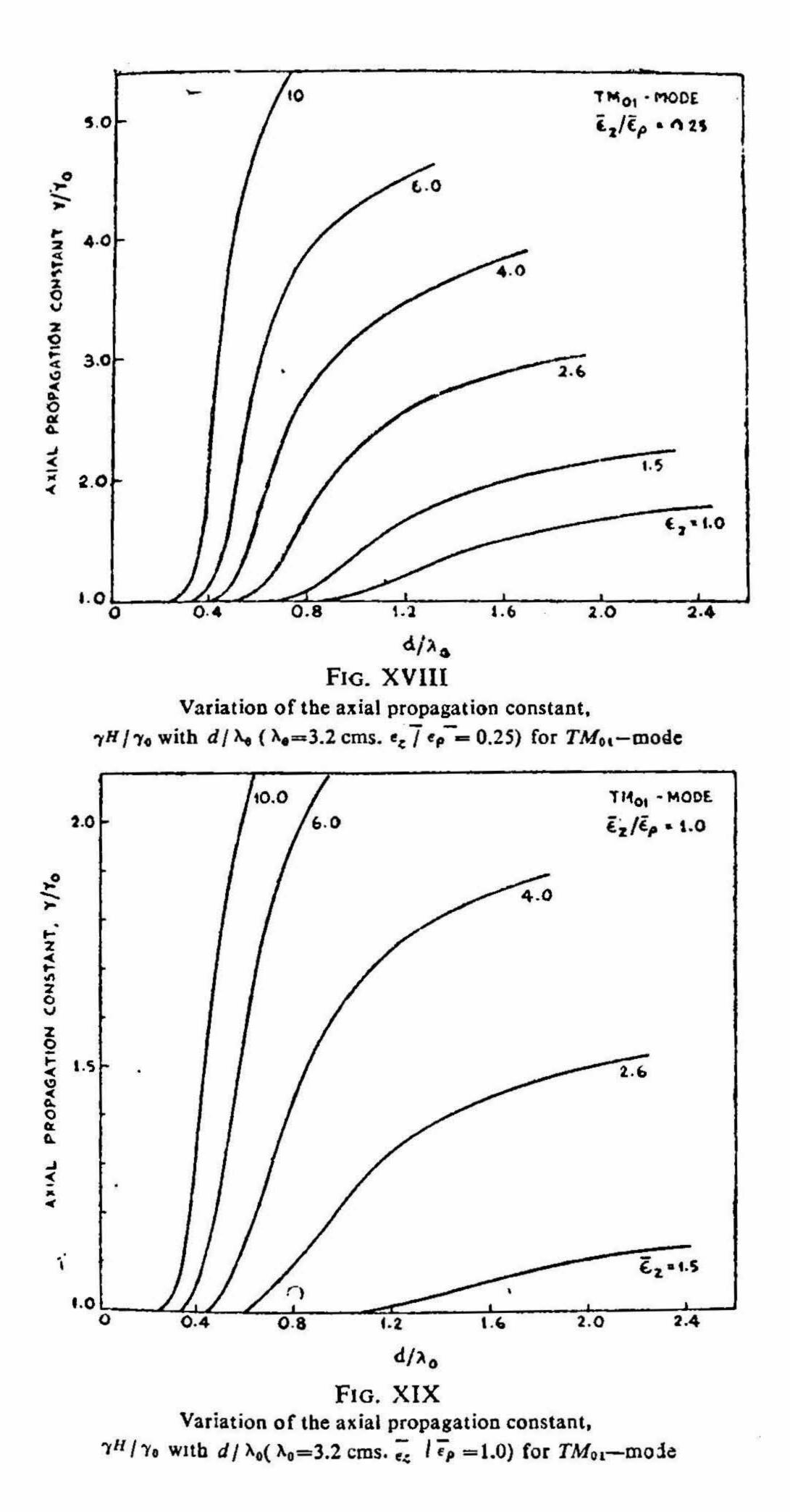


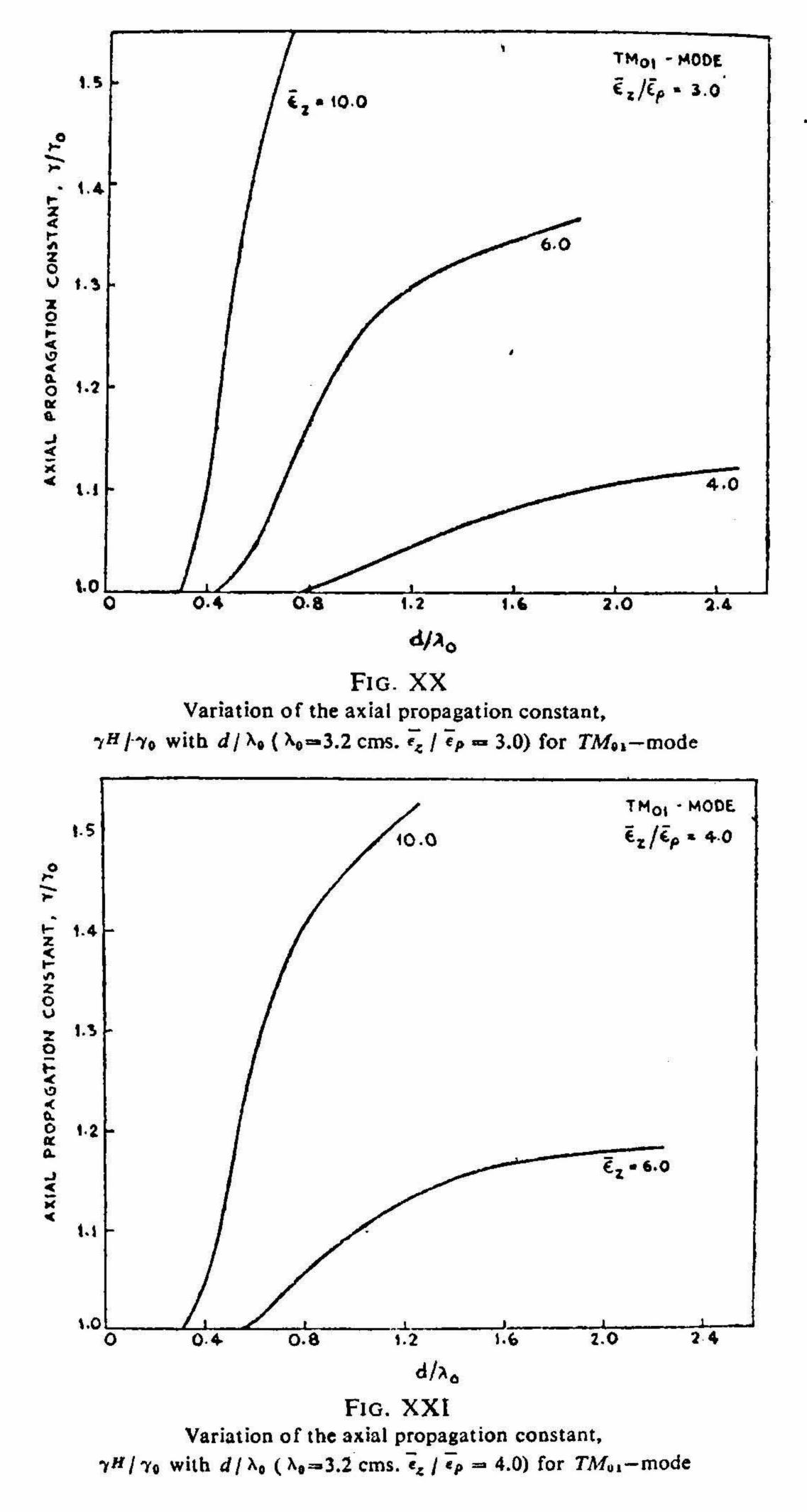
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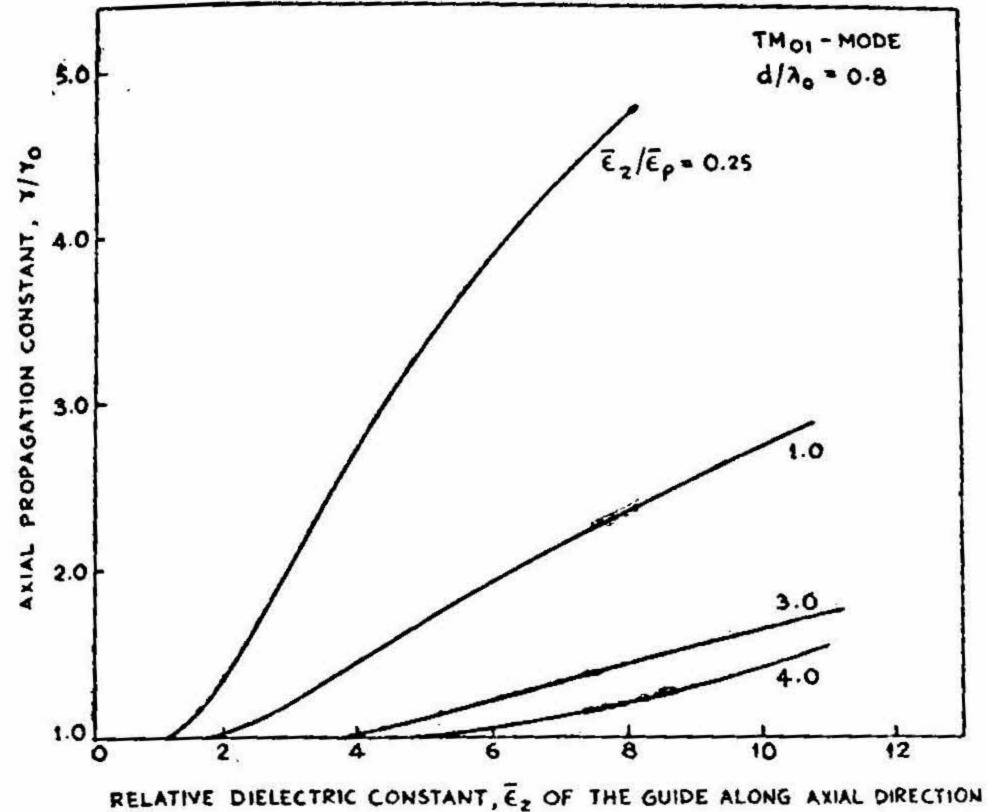
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1.1.1







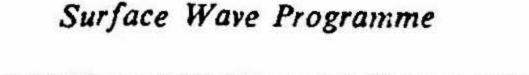


FIG. XXII

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Variation of the axial propagation constant,

 $\gamma H / \gamma_0$ with ϵ_z (d/ $\lambda_0 = 0.8$, $\lambda_0 = 3.2$ cms.) for TM_{01} -mode

4.5 Cut-off Conditions

4.

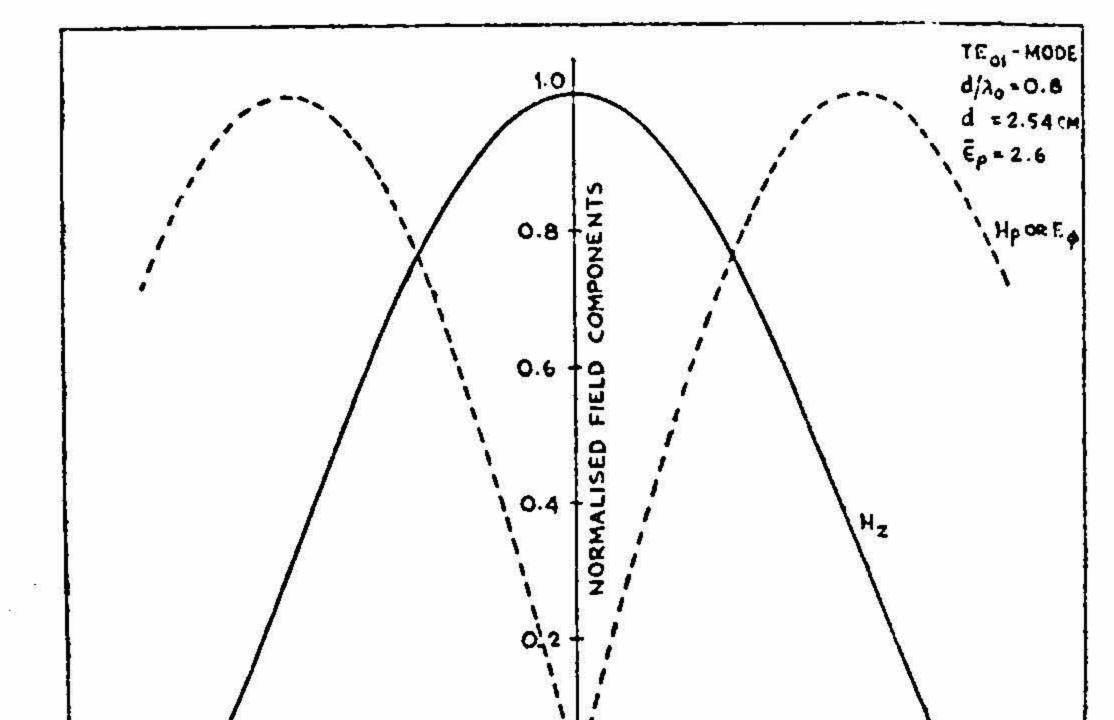
It is seen from the solutions of the characteristic equations [10], [12], [13] and [14] that for the axially symmetric modes, propagation along the guide is not possible without radial loss of energy for waveguide diameters less than a limiting diameter and for wavelengths greater than a corresponding limiting wavelength depending on the dielectric constants and the permittivities of the medium. This is the cut-off phenomenon for these modes. Immediately above this limiting value, $\lambda_g \approx \lambda_0$ and only a smaller part of the energy flows inside the guide.

For values of (d/λ_0) greater than the limiting value, the guided wavelength becomes smaller and smaller and approaches $[\lambda_0 (\bar{\epsilon}_p)^{-1}]$ and power flow tends rapidly to be confined totally within the rod. Or, in other words, at $(\lambda_g/\lambda_0) \ge 1$ and with real arguments of Hankel functions, solution of the characteristic equation yields a minimum value of (d/λ_0) , which we define as the condition of cut-off. They are given as,

$$\frac{(d/\lambda_0) \operatorname{cut-off}}{\pi} \left[\frac{S_n}{\pi} \left[\frac{-\epsilon_z}{\epsilon_z} - \left(\frac{-\epsilon_z}{\epsilon_p} \right) \right]^{-1} \text{ for } TM_{0n} \text{ waves} \right]$$

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where, S_n = value of the argument of $J_0(x)$ for $J'_0(x) = 0$ and, n = 1, 2, 3, 4... indicating dominant and higher order mode configurations.



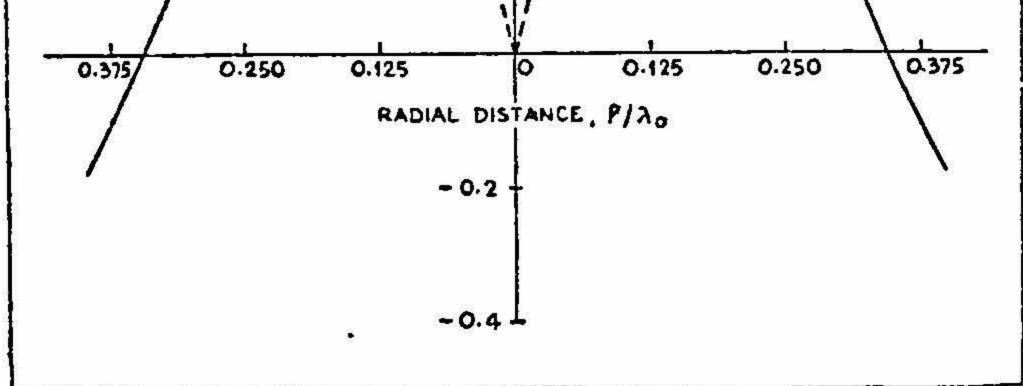
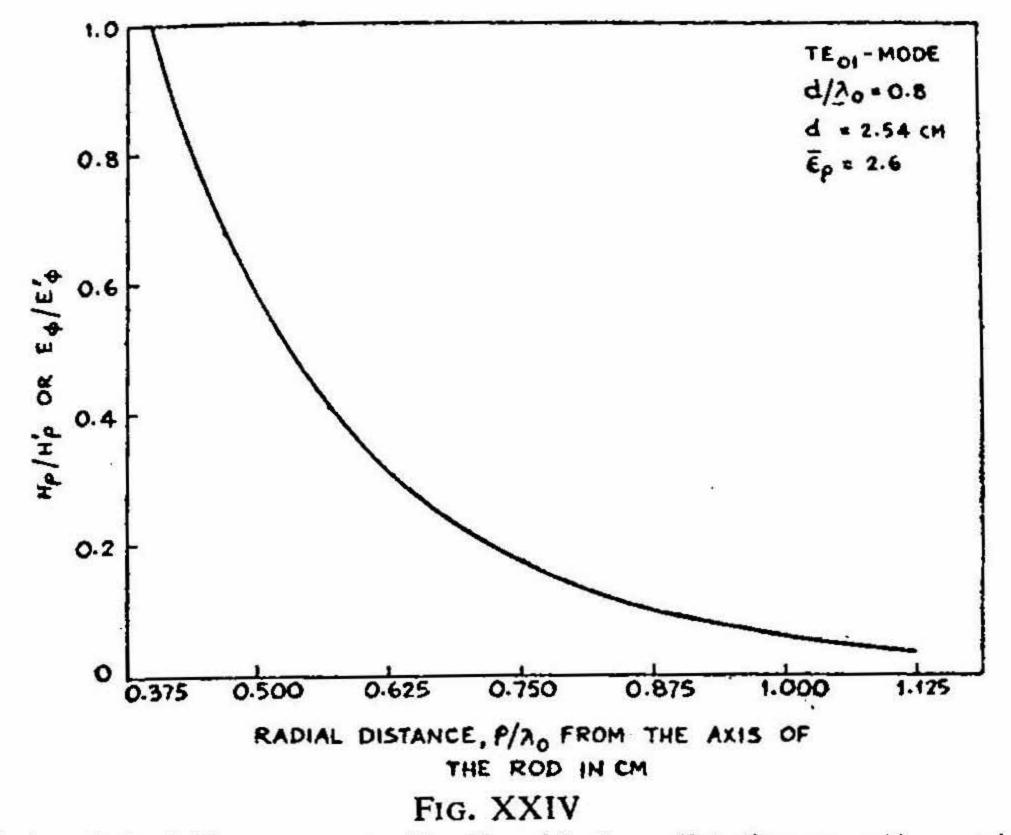
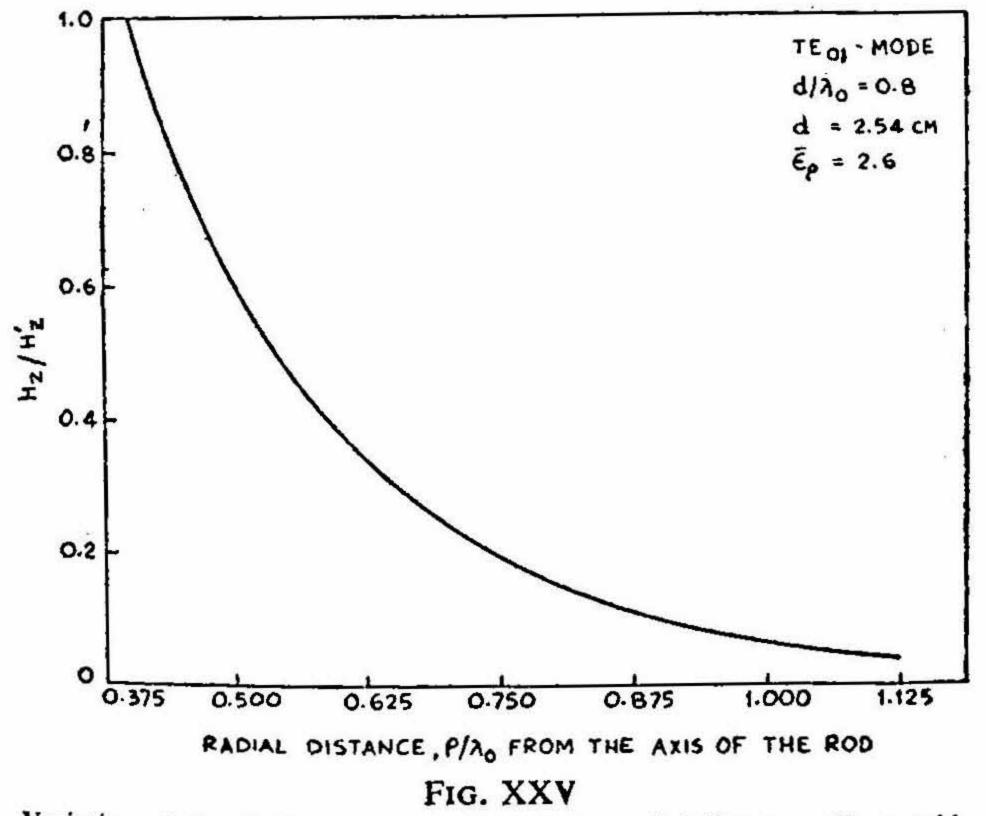


FIG. XXIII

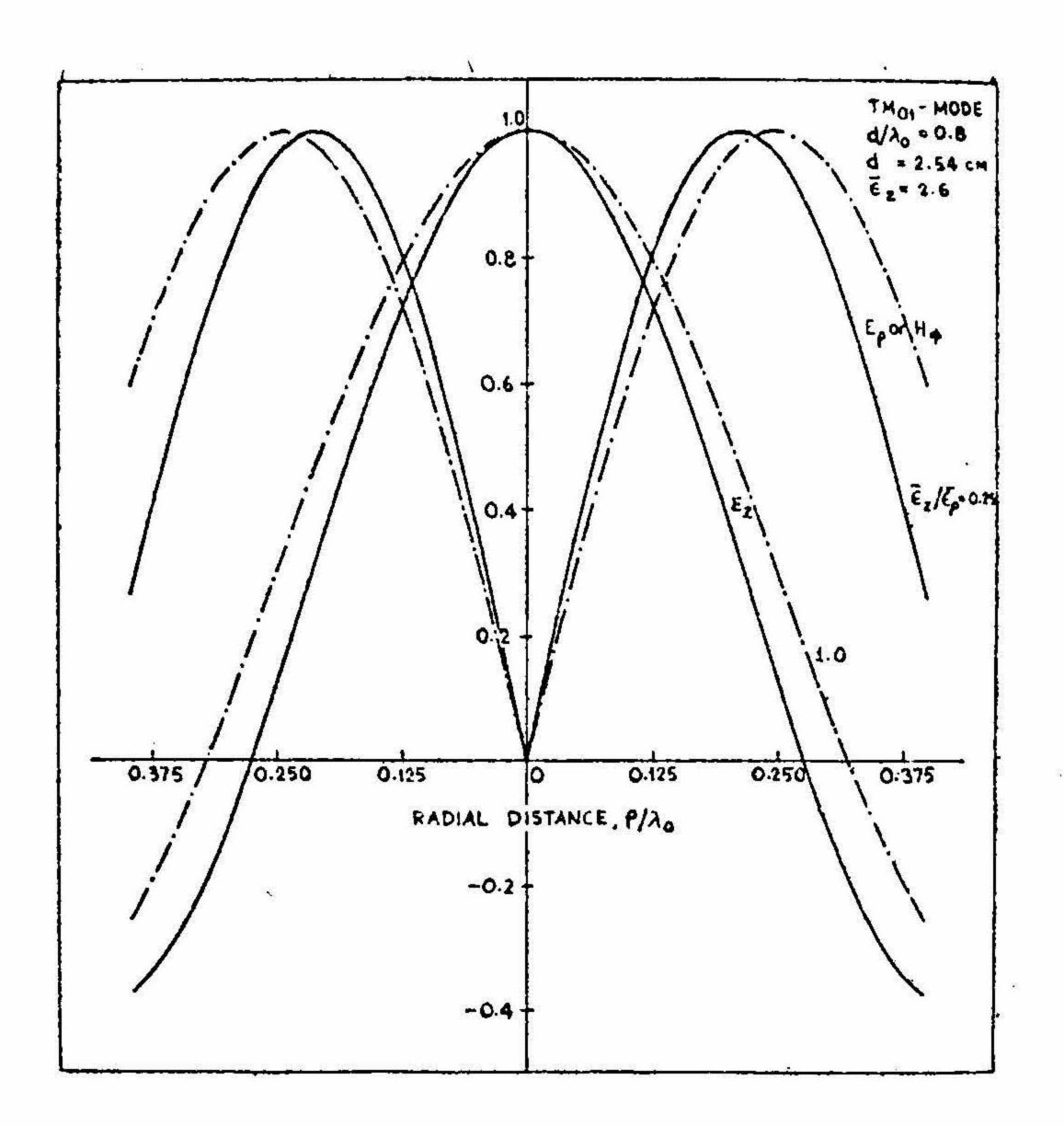
Variation of the field components, H_z , E_ϕ , H_ρ , with the radial distance, ρ/λ_0 , inside the guide for TE_{01} -mode. ($\overline{\epsilon_{\rho}}=2.6$, $d/\lambda_0=0.8$, $\lambda_0=3.2$ cms.) (E_ϕ , H^{ρ} : have been normalised with respect to their values E'_{ϕ} and H'_{ρ} , respectively, at $\rho=0.8$ cms.)



Variation of the field components, H_{ρ} , H_{ϕ} with the radial distance, ρ / λ_0 , outside the guide for TE_{01} mode ($\epsilon_{\rho} = 2.6$, $d/\lambda_0 = 0.8$, $\lambda_0 = 3.2$ cms.) (H_{ρ} , E_{ϕ} have been normalised with respect to their vales H'_{ρ} and E'_{ϕ} , respectively on the surface of waveguide, $\rho = r$)



Variation of the field component H_z with the radial distance ρ / λ_0 outside guide for TE_{01} mode ($\overline{e}\rho = 2.6$, $d/\lambda_0 = 0.8$, $\lambda_0 = 3.2$ cms.) (H_z has been normalised with respect to its value. H'_z at the surface of the waveguide, $\rho = r$)



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FIG. XXVI

Tariation of the field components, E_z , E_ρ , H_ϕ with the radial distance, P/λ_0 , inside the guide for TM_{01} mode ($\overline{\epsilon_z} = 2.6, d/\lambda_0 = 0.8, \lambda_0 = 3.2$ cms: $\overline{\epsilon_z}/\overline{\epsilon_p} = 2.5, 1.0$) (E_ρ , H_ϕ have been normalised with respect to their values, E'_ρ and H'_ϕ respectively, at P = 0.675 cms, 0.7853 cms. for the two different $\overline{\epsilon_z}/\overline{\epsilon_p}$.)

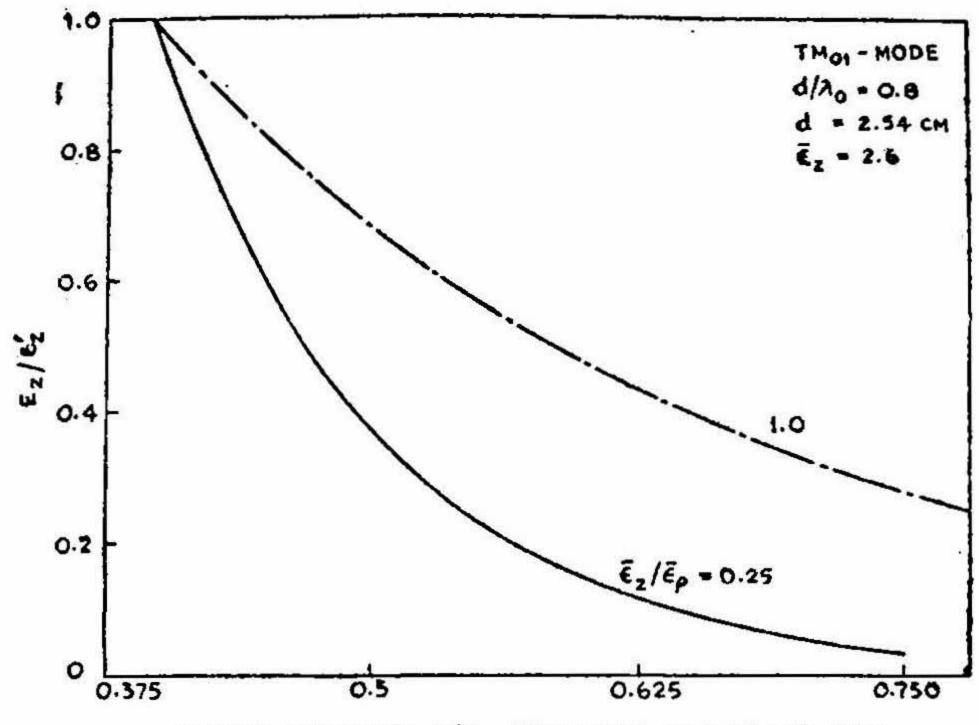




FIG. XXVII

Variation of the field component, E_z with the radial distance, P/λ_0 outside the guide for TM_{01} —mode ($\overline{e_z} = 2.6$, $d/\lambda_0 = 0.8$, $\lambda_0 = 3.2$ cms, $\overline{e_z}/\overline{e_p} = 0.25$, 1.0) E_z has been normalised with respect to its value E'_z on the surface of the waveguide, P=r).

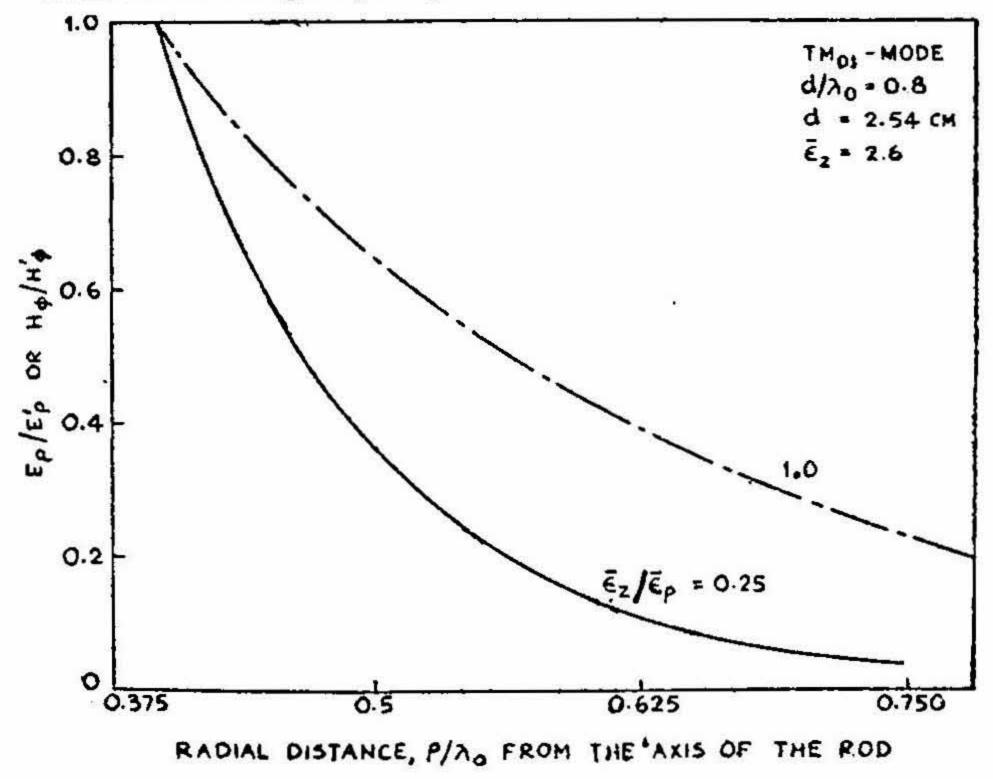


FIG XXVIII

Variation of the field components, E_{ρ} , H_{ϕ} with the radial distance, ρ/λ_{0} , outside the guide for TM_{01} -mode ($\overline{\epsilon_{z}} = 2.6$, $d/\lambda_{0} = 0.8$, $\lambda_{0} = 3.2$ cms, $\overline{\epsilon_{z}}/\overline{\epsilon_{\rho}} = 0.25$, 1.0 (E_{ρ} , H_{ϕ} have been normalised with respect to their values E'_{ρ} , H'_{ϕ} , respectively on the surface of the guide, $\rho = r$ for the two different $\epsilon_{z}/\overline{\epsilon_{\rho}}$). For dominant modes, $S_n = 2.405$ and cut-off conditions are plotted as a function of ϵ_p for TE_{01} mode and as a function of ϵ_2 with $[\epsilon_z/\epsilon_p]$ ranging from 0.25 to 4.0 for TM_{01} mode in the Figures XXIX and XXX, respectively.

5. CONCLUSION

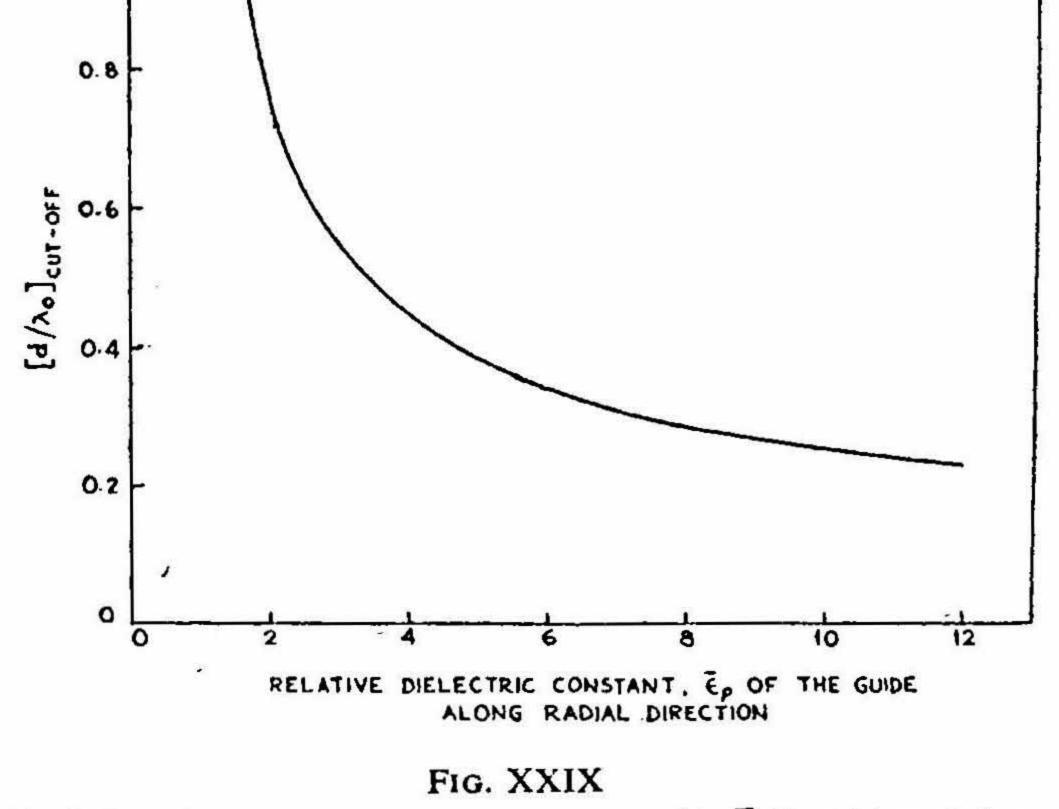
It is observed from the graphs of the guided-wavelength that λ_g decreases with an increase in the dielectric constants, $\overline{\epsilon_z}$, $\overline{\epsilon_p}$. The decrease of λ_g with an increase in the diameter of the guide may be due to the increase in the axial propagation constant, as an increase in $\gamma = i\beta$, for a loss-less waveguide, reduces the phase velocity of the propagating wave. In the case of *TM*waves, λ_g increases with an increase of $[\overline{\epsilon_z}/\overline{\epsilon_p}]$ for any $\overline{\epsilon_z}$.

From the curves of radial and axial propagation constants it will be seen that K_1 decreases and K_2 and γ increase with larger diameter of the guide, and with an increase in the diclectric constants, $\overline{\epsilon_2}$, $\overline{\epsilon_p}$ all the three constants, K_1 , K_2 and γ increase. However, the rate of increase of K_1 with increase of dielectric constants is less compared to those of K_2 and γ . This accounts for the reduction in the radial-field spread for such wave-guides having larger diameters and higher dielectric constants.

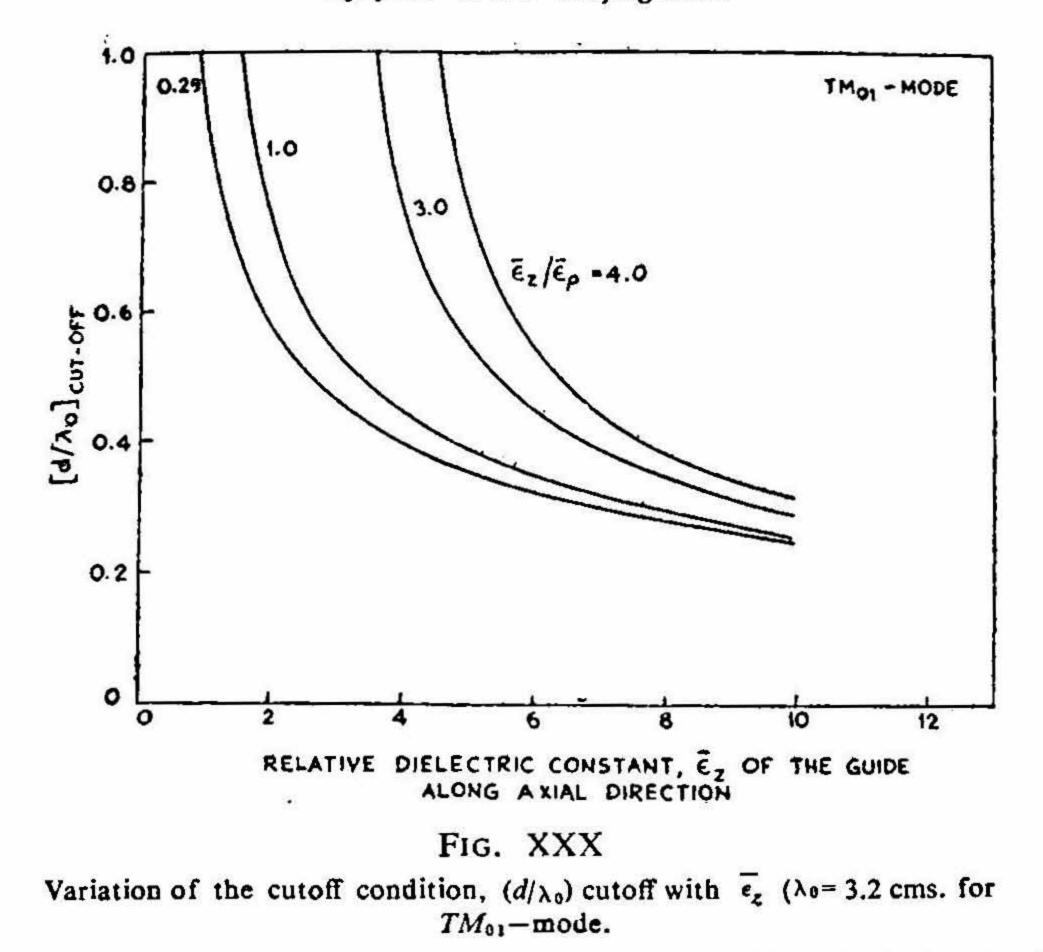
It will be evident from equation (3) that on substituting $[\epsilon_z/\epsilon_\rho] = 1$, the propagation characteristics for TM_{01} mode in an isotropic dielectric rod waveguide (Kiely, 1953) are obtained. For the TE_{01} mode, however, the solutions

TE OI - MODE

1.0



Variation of the cutoff condition, (d/λ_0) cutoff with $\overline{e_p}$ ($\lambda_0 = 3.2$ cms.) for TE_{01} -mode.



are similar to those for the isotropic wave-guides as $\overline{\epsilon}_{\rho}$ only is involved due to the nature of the field configurations.

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For the circularly symmetrical TM_{01} mode and $(d/\lambda_0) = 0.8$, $\overline{\epsilon_z} = 6.0$, the propagation characteristics of surface waves through an isotropic dielectric rod waveguide have been compared with those for an isotropic dielectric rod guide in Table I.

$\left[\overline{\epsilon}_{z}/\overline{\epsilon}_{\rho}\right]$	0.25	1.0	3.0	4.0
$[\lambda_{g}/\lambda_{0}]$	0.258	0.511	0 810	0.943
<i>K</i> ^{<i>H</i>} ₁	2.95	2 89	2.71	2.47
K_2^H/i	7.39	3.32	1.37	0.697
[7/70]	3 83	1.93	1.22	1.04
$[d/\lambda_0]$ cutoff	0 320	0.347	0.443	0.544

TABLE I

Comparative Study of the Propagation Characteristics

From the above it is quite clear that for any anisotropic dielectric rod waveguide, the fields get more and more concentrated inside the guide with decreasing anisotropicity, *i.e.*, $\left[\frac{\overline{\epsilon_{z}}}{\overline{\epsilon_{p}}}\right]$ tending to lower values. This results into the relative advantages over its isotropic counterpart in transmitting power over a longer distance along the guide.

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LIST OF SYMBOLS

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E	Electric field vector.		
H	Magnetic field vector.		
μ	Permeability of the outer medium, air.		
E	Dielectric constant of the outer medium, air.		
<i>ϵ</i> 1	Dielectric constant of the anisotropic dielectric rod.		

Surface Wave Propagation

€p	Tensor component of ϵ_1 in the radial direction.
€z	Tensor component of ϵ_1 in the longitudinal direction.
γ_1^E, γ_2^E	Axial propagation constants inside and outside the wave- guide respectively, for TE_{01} mode.
γ_1^H, γ_2^H	Axial propagation constants inside and outside the waveguide respectively, for TM_{01} mode.
λ_{g}	Guided wave-length, cm.
λο	Wavelength in free-space, 3.2 cm.
d	Diameter of the waveguide, cm.
K_{1}^{E}, K_{2}^{E}	Radial propagation constants inside and outside the wave- guide respectively, for TE_{01} mode.
K_1^H, K_2^H	Radial propagation constants inside and outside the wave- guide respectively, for TM_{01} mode.
A, B, C, D	Constants to be determined from boundary conditions.
$J_0(x), H_0^{(1)}(x)$	Zero-order Bessel and Hankel functions of the first kind, respectively.
$J_{1}(x), H_{1}^{(1)}(x)$	First-order Bessel and Hankel functions of the first kind, respectively.